

# On Surrogate Channels for Code Design

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# Outline

- **Popular Surrogate Channels:** BEC and biAWGN
- **Design Challenge:** Polar Codes for 8-ASK
- First Attempt
- Second Attempt
- **Detour:** Mismatched Decoding
- Final Attempt
- Conclusions

# BEC and biAWGN

- **Idea:** analyze decoding by tracking mutual information (MI) as it propagates through factor graph.
- Two channels are “**easy**”:
  - Binary Erasure Channel (BEC)
  - biAWGN Channel (with Gaussian Approximation)<sup>12</sup>

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<sup>1</sup>S. ten Brink, G. Kramer, and A. Ashikhmin, “Design of low-density parity-check codes for modulation and detection,” 2004.

<sup>2</sup>F. Brännström, “Convergence analysis and design of multiple concatenated codes,” Ph.D. dissertation, 2004.

# Code Design by Surrogate Channels

- **Observation:** codes often have “universal” properties.<sup>3</sup>
- **Idea:** Design code for surrogate channel, use it for target channel.<sup>4</sup>
- **Higher-order modulation:** use surrogate channel for each bit-channel.<sup>5</sup>

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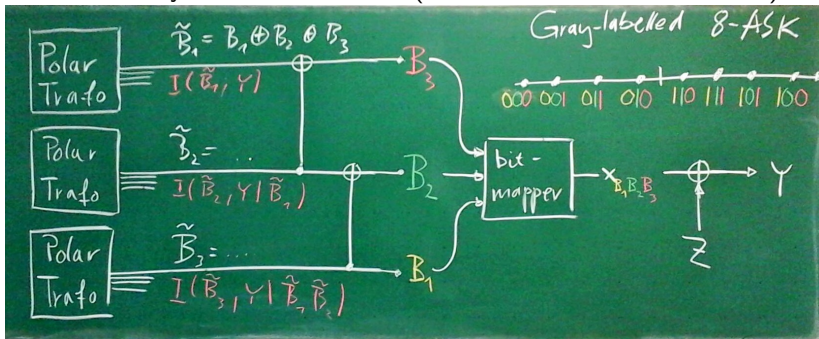
<sup>3</sup>M. Franceschini, G. Ferrari, and R. Raheli, “Does the performance of LDPC codes depend on the channel?” 2006.

<sup>4</sup>F. Peng, W. E. Ryan, and R. D. Wesel, “Surrogate-channel design of universal LDPC codes,” 2006.

<sup>5</sup>F. Steiner, G. Böcherer, and G. Liva, “Protograph-based LDPC code design for shaped bit-metric decoding,” 2016.

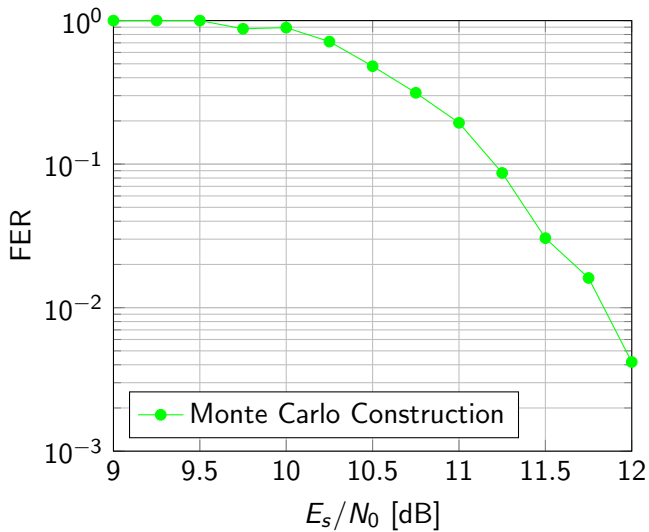
## Design Challenge: Polar Code for 8-ASK<sup>6</sup>

For the following modulation scheme, we want to **efficiently** construct Polar codes that perform as good as polar codes constructed by exhaustive search (“Monte Carlo Construction”).



<sup>6</sup>H. MahdaviFar, M. El-Khomy, J. Lee, and I. Kang, “Polar coding for bit-interleaved coded modulation,” 2016.

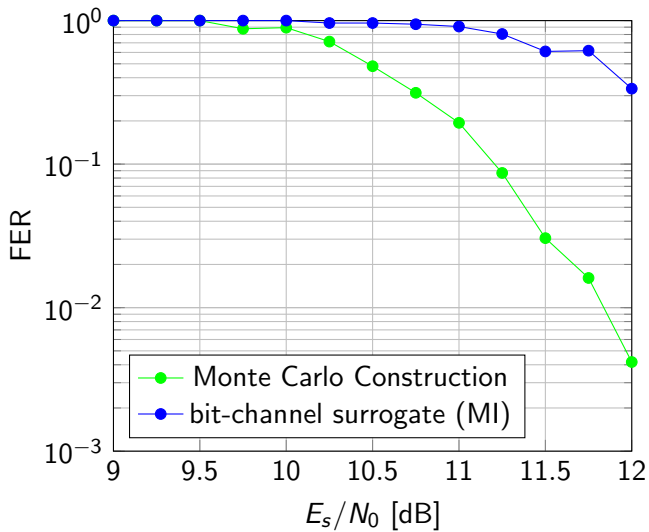
1.5 bpcu, 8-ASK,  $c = 1/2$ ,  $n = 512$



# First Attempt

- Three bit-channels  $p_{Y|B_i}$ ,  $i = 1, 2, 3$ .
- Replace the  $i$ th bit-channel by a biAWGN surrogate channel with MI equal to  $I(B_i; Y)$ ,  $i = 1, 2, 3$ .

1.5 bpcu, 8-ASK,  $c = 1/2$ ,  $n = 512$





## Second Attempt

- Replace the polarized bit-channels by biAWGN surrogate channels with MI equal to

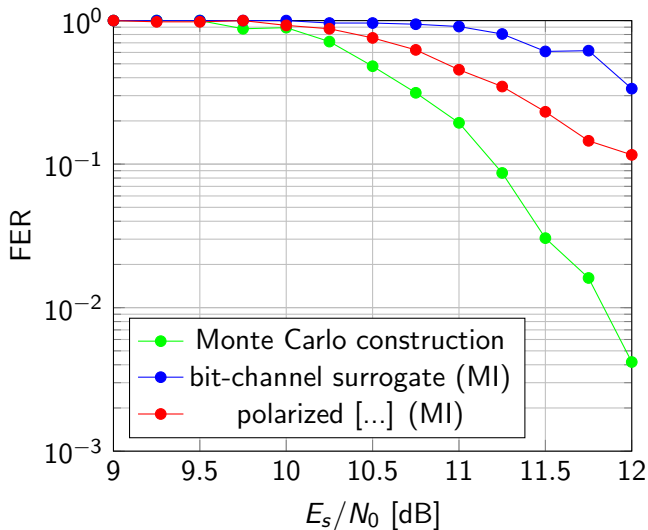
$$I(\tilde{B}_1; Y)$$

$$I(\tilde{B}_2; Y | \tilde{B}_1)$$

$$I(\tilde{B}_3; Y | \tilde{B}_1 \tilde{B}_2)$$

respectively.

1.5 bpcu, 8-ASK,  $c = 1/2$ ,  $n = 512$



## Detour: Mismatched Decoding

### What is mutual information?

- Memoryless channel  $p_{Y|X}$ .
- Random codebook  $\mathcal{C} = \{X^n(1), \dots, X^n(2^{nR})\}$  with entries iid  $P_X$ .
- Message  $w \in \{1, 2, \dots, 2^{nR}\}$
- ML decoder

$$\hat{W} = \underset{w \in \{1, 2, \dots, 2^{nR}\}}{\operatorname{argmax}} p_{Y^n|X^n}(Y^n|X^n(w)) = \prod_{i=1}^n p_{Y|X}(Y_i|X_i(w))$$

- Error probability  $\Pr(W \neq \hat{W}) \xrightarrow{n \rightarrow \infty} 0$  if

$$R < I(X; Y).$$

## Detour: Mismatched Decoding<sup>78</sup>

- (Random coding as on previous slide).
- Auxiliary channel  $q(\cdot|\cdot)$
- Mismatched Decoder

$$\hat{W} = \underset{w \in \{1, 2, \dots, 2^{nR}\}}{\operatorname{argmax}} \prod_{i=1}^n q_{Y|X}(Y_i|X_i(w))$$

- Achievable rate

$$R_{\text{LM}} = \max_{s,r} \mathbb{E} \left[ \log \frac{q(Y|X)^s r(X)}{q_{s,r}(Y)} \right]$$

- Auxiliary output distribution  $q_{s,r}(\cdot) = E[q(\cdot|X)^s r(X)]$
- $s \geq 0$ , function  $r: \mathcal{X} \rightarrow \mathbf{R}$ .

<sup>7</sup>A. Ganti, A. Lapidoth, and E. Telatar, “Mismatched decoding revisited: General alphabets, channels with memory, and the wide-band limit,” 2000.

<sup>8</sup>G. Böcherer, “Achievable rates for shaped bit-metric decoding,” 2016.

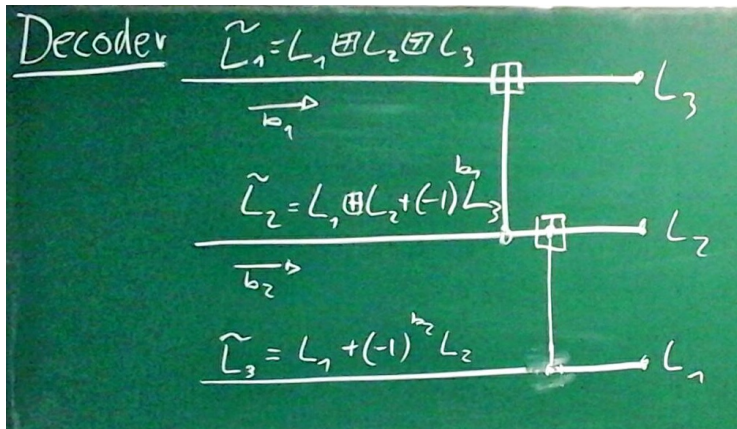
## Detour: Mismatched Decoding

- Let's account for what the decoder is actually doing!
- $L$ -value defines auxiliary channel via

$$L = \log \frac{q_{B|Y}(0|y)}{q_{B|Y}(1|y)}$$
$$q_{B|Y}(b|y) = \frac{e^{-L \cdot b}}{1 + e^{-L}}$$

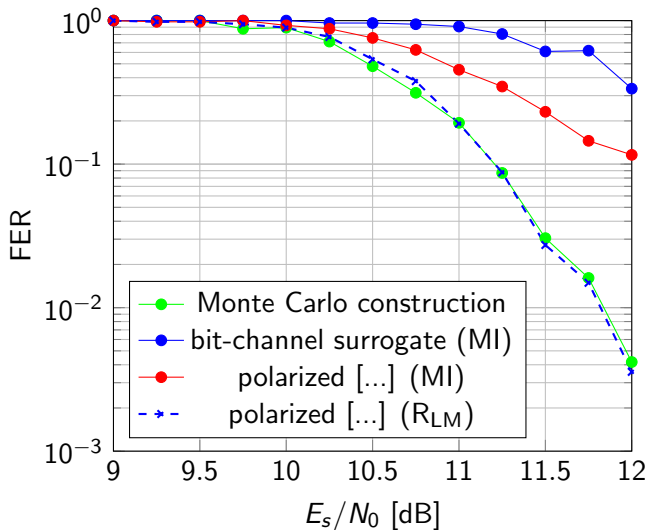
⇒ Estimate  $R_{LM}$  for  $L$ -value as decoding metric.

## Final Attempt



- Replace each polarized bit-channel by a biAWGN surrogate channel with MI matched to  $R_{LM}$  estimated from  $\tilde{L}_i, i = 1, 2, 3$ .

1.5 bpcu, 8-ASK,  $c = 1/2$ ,  $n = 512$



## Conclusions

- Monte Carlo construction: half an hour (implementation in C).
- Surrogate channel design: some milliseconds.
- Construction with appropriate surrogate channel yields performance as good as Monte Carlo construction.
- Tool: information theory for mismatched decoding.



# Literature

- [1] S. ten Brink, G. Kramer, and A. Ashikhmin, "Design of low-density parity-check codes for modulation and detection," *IEEE Trans. Commun.*, vol. 52, no. 4, pp. 670–678, 2004.
- [2] F. Brännström, "Convergence analysis and design of multiple concatenated codes," Ph.D. dissertation, Chalmers University of Technology, 2004.
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- [4] F. Peng, W. E. Ryan, and R. D. Wesel, "Surrogate-channel design of universal LDPC codes," *IEEE Commun. Lett.*, vol. 10, no. 6, pp. 480–482, Jun. 2006.
- [5] F. Steiner, G. Böcherer, and G. Liva, "Protograph-based LDPC code design for shaped bit-metric decoding," *IEEE J. Sel. Areas Commun.*, vol. 34, no. 2, pp. 397–407, Feb. 2016.
- [6] H. Mahdaviifar, M. El-Khamy, J. Lee, and I. Kang, "Polar coding for bit-interleaved coded modulation," *IEEE Trans. Veh. Technol.*, vol. 65, no. 5, pp. 3115–3127, May 2016.
- [7] A. Ganti, A. Lapidot, and E. Telatar, "Mismatched decoding revisited: General alphabets, channels with memory, and the wide-band limit," *IEEE Trans. Inf. Theory*, vol. 46, no. 7, pp. 2315–2328, Nov. 2000.
- [8] G. Böcherer, "Achievable rates for shaped bit-metric decoding," *arXiv preprint*, 2016. [Online]. Available: <http://arxiv.org/abs/1410.8075>