

# Three Open Problems in Probabilistic Shaping

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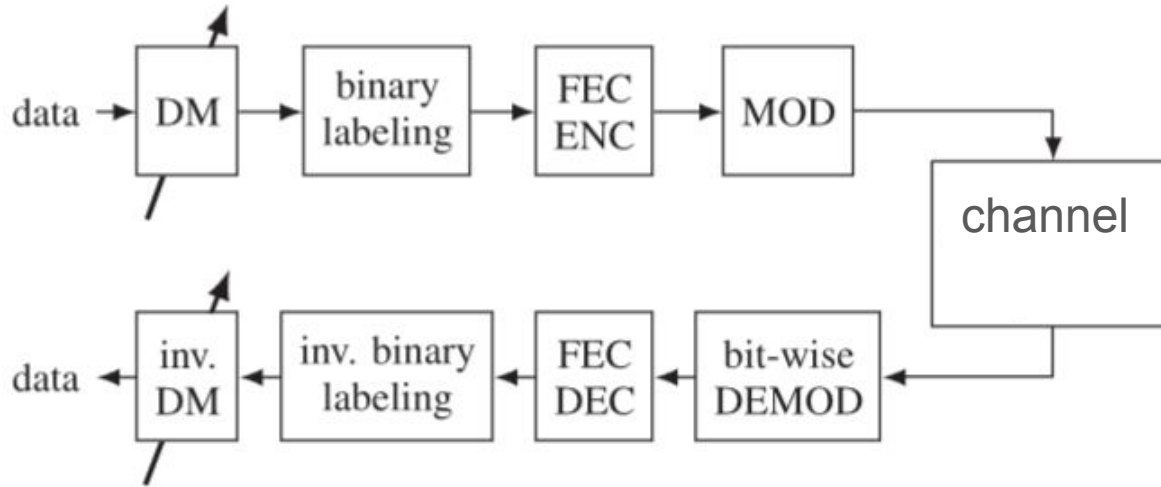
# Outline

- Probabilistic Amplitude Shaping
- An Optimization Problem
- An Information Theoretic Problem
- A Machine Learning Problem

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# Probabilistic Amplitude Shaping

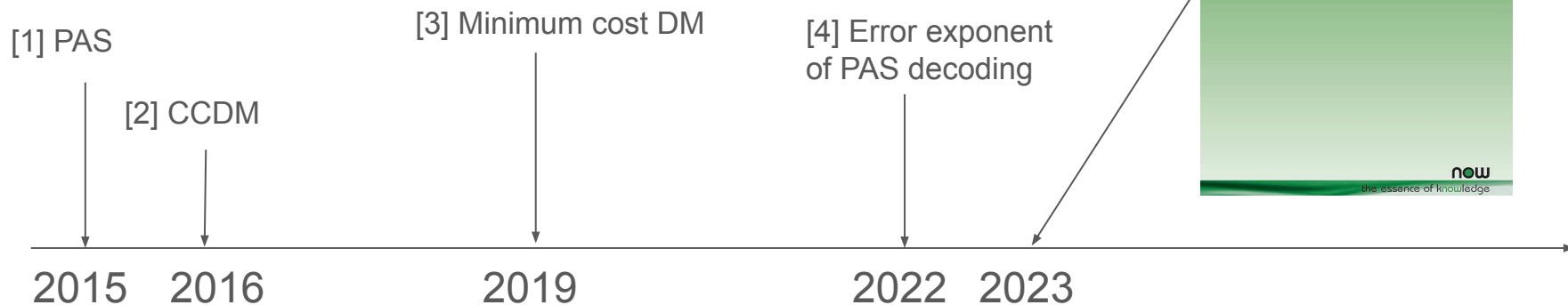


- **PAS encoding:** systematic linear FEC encoder preserves amplitude distribution **imposed by DM**.
- **PAS decoding:** linear FEC decoder on full linear code, **agnostic of DM**.

**This talk focuses on theoretic aspects of PAS**

# Probabilistic Amplitude Shaping

## *Theoretical advances*



[1] G. Böcherer, F. Steiner, and P. Schulte, “Bandwidth efficient and rate-matched low-density parity-check coded modulation,” *IEEE Trans. Commun.*, vol. 63, no. 12, pp. 4651–4665, Dec. 2015

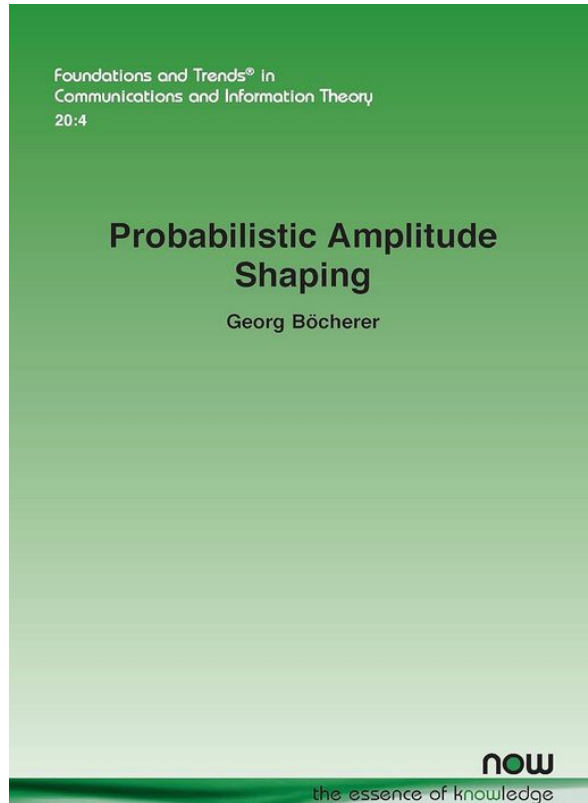
[2] P. Schulte and G. Böcherer, “Constant composition distribution matching,” *IEEE Trans. Inf. Theory*, vol. 62, no. 1, pp. 430–434, Jan. 2016

[3] P. Schulte and F. Steiner, “Divergence-optimal fixed-to-fixed length distribution matching with shell mapping,” *IEEE Wireless Commun. Letters*, vol. 8, no. 2, pp. 620–623, Apr. 2019

[4] N. Merhav and G. Böcherer, “Codebook mismatch can be fully compensated by mismatched decoding,” *IEEE Trans. Inf. Theory*, vol. 69, no. 4, pp. 2152–2164, Apr. 2023

# Probabilistic Amplitude Shaping

<https://github.com/gbsha/PAS>



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## Setup

- ▶  $2^m$ -ASK constellation

$$X \in \{\pm 1, \pm 3, \dots, \pm(2^m - 1)\}$$

- ▶ AWGN channel

$$Y = X + \sigma Z, \quad Z \sim \mathcal{N}(0, 1)$$

- ▶ Signal-to-noise ratio

$$\text{SNR} = \frac{\mathbb{E}(X^2)}{\sigma^2}$$

- ▶ FEC rate

$$0 \leq R_{\text{fec}} \leq 1$$



## PAS Achievable Rate

- ▶ Maxwell-Boltzmann (MB) distribution

$$P_X(x) \propto \exp(-\nu|x|^2)$$

- ▶ Spectral efficiency

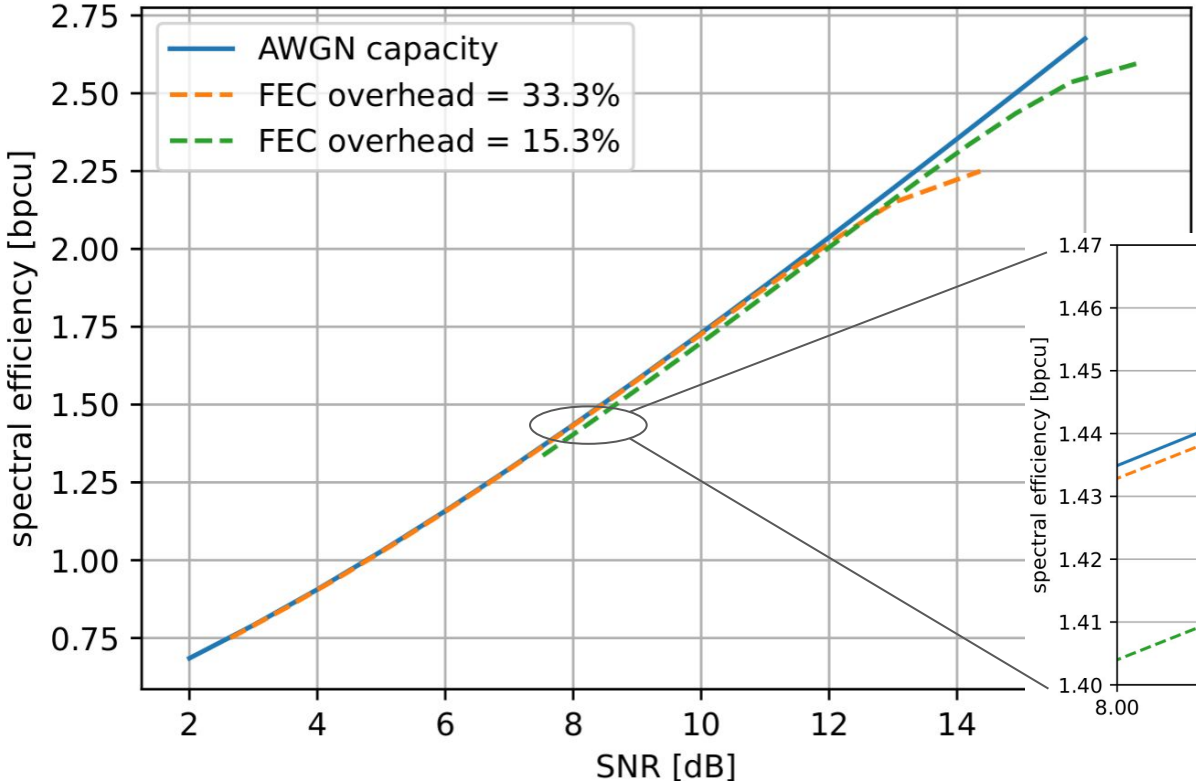
$$\text{SE} = \mathbb{H}(X) - m(1 - R_{\text{fec}})$$

- ▶ By [5, Example 5.5], an achievable noise level is

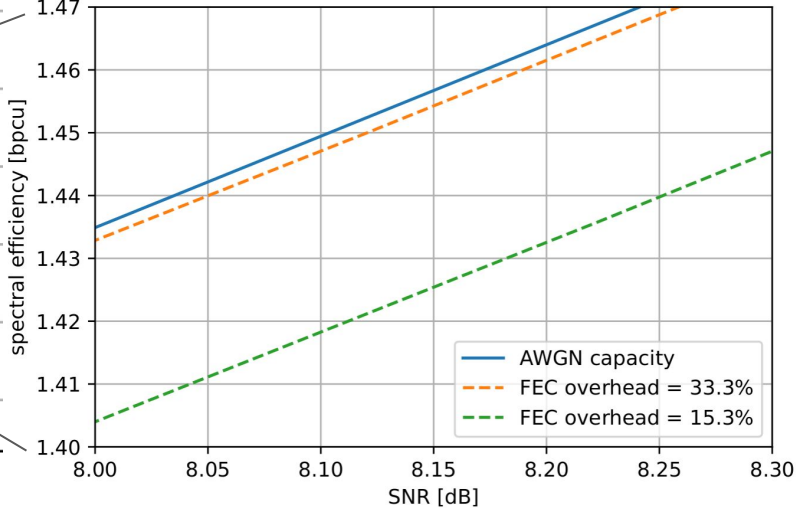
$$\sigma: \mathbb{H}(X|X + \sigma Z) = m(1 - R_{\text{fec}})$$

Killer application in optical communications:

# PAS Achievable Rate



- Operate over a large range of SNRs,
- close to capacity,
- with a fixed, low overhead FEC engine



## [5, Sec. 1.4.5]

What if we want to optimize  $P_X$ ?

$$\begin{aligned} & \underset{P_X}{\text{maximize}} && \mathbb{H}(X) \\ & \text{subject to} && \mathbb{H}(X|Y) \leq m(1 - R_{\text{fec}}) \end{aligned}$$

The Lagrangian is

$$L(P_X, \lambda) = \mathbb{H}(X) - \lambda \mathbb{H}(X|Y)$$

- ▶ For  $\lambda = 1$ :  $L(P_X, \lambda) = \mathbb{I}(X; Y) \Rightarrow$  convex  $\cap$  in  $P_X$ .
- ▶ For large enough  $\lambda > 1$ ,  $L(P_X, \lambda)$  becomes **non-convex**

**Problem 1:** Find algorithm to calculate optimal  $P_X$

## Remarks on Problem 1

- ▶ The solution may assign probability 0 to some signal points.
- ▶ Enhancing the peak power constrained problem in [6] by a constraint on the equivocation may be a viable approach.

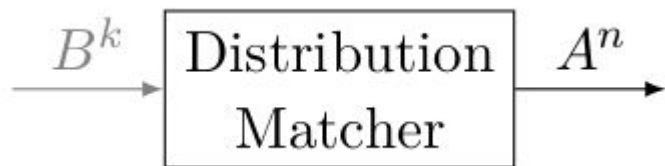
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[6] J. G. Smith, "The information capacity of amplitude-and variance-constrained scalar Gaussian channels," *Inf. control*, vol. 18, no. 3, pp. 203–219, 1971

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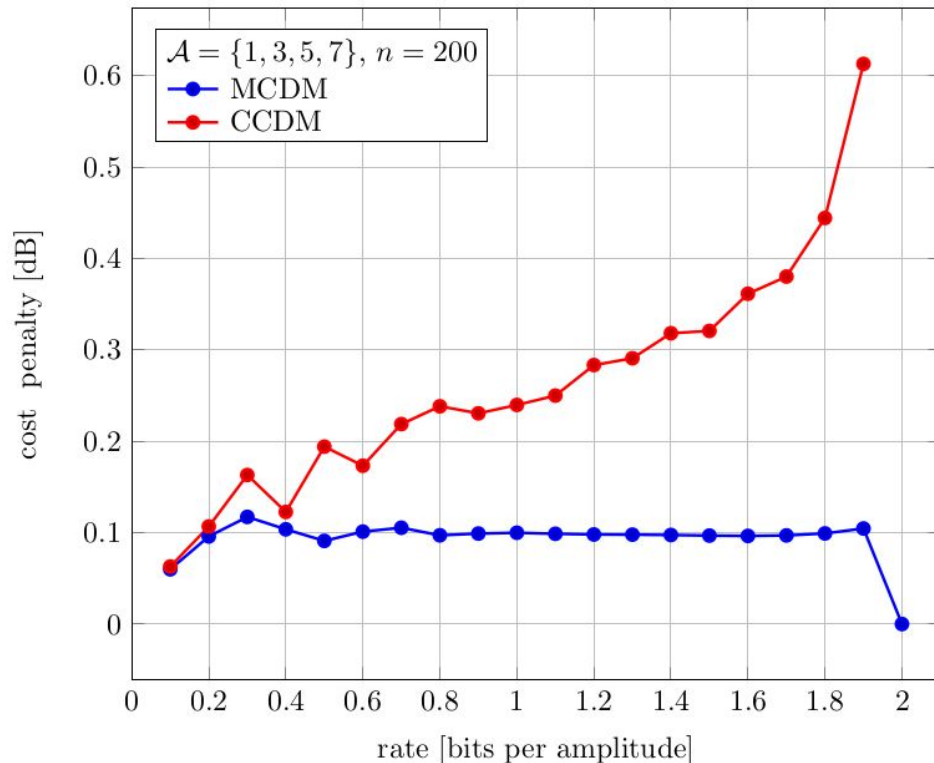
# Distribution Matching [5, Chapter 2]



- ▶ Rate  $R = k/n$
- ▶ Cost  $\mathbb{E}[w(\bar{A})]$
- ▶ MB source  $A^*$  with  $\mathbb{H}(A^*) = R$

$$\text{cost penalty} = 10 \log_{10} \frac{\mathbb{E}[w(\bar{A})]}{\mathbb{E}[w(A^*)]}$$

**Remark:** The CCDM distribution (“type”) is chosen in [5] according to the minimum cost for which the required rate is achieved. This results in lower cost than starting from an MB distribution and quantizing it to an n-type. The minimum cost DM (MCDM) indexes the  $2^{nk}$  least cost length n sequences. Thus, both for CCDM and MCDM, the distribution (averaged both over the sequences and the entries of the sequences) has only a subordinate role. Further, in practice, the cost penalty has proven to be a more relevant metric for DM comparison than rate loss or relative entropy, for two reasons: (1) the MB distribution itself results from minimizing cost; so directly minimizing cost is more meaningful than approximating an MB distribution. (2) rate is a system parameter, which usually has to be realized exactly.



## PAS Channel Coding Theorem

- ▶ By [5, Theorem 5.3], asymptotically in FEC length  $n$ , PAS can achieve

$$SE^* = R_{\text{ccdm}}(P_{|X|}, n_{\text{dm}}) + 1 - \mathbb{H}(X|Y)$$

with  $\text{sign}(X)$  uniform on  $\{-1, 1\}$ .

- ▶ For  $n_{\text{dm}} \rightarrow \infty$ ,

$$SE^* = \mathbb{H}(X) - \mathbb{H}(X|Y)$$

which is the best we can do.

- ▶ However, for short DM length, we know empirically that MCDM is much better than CCDM. **Example:** 800G: FEC decoding window  $> 100\,000$ ,  $n_{\text{dm}} = 12$ .

**Problem 2:** Prove PAS channel coding theorem for MCDM

## Comments on Problem 2

- ▶ [4] proves that PAS decoding of linear codes achieves the best known error exponent for constant composition codes [7, Theorem 10.2].
- ▶ [8] proves a PAS channel coding theorem where the DM is replaced by a discrete memoryless source.

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[4] N. Merhav and G. Böcherer, "Codebook mismatch can be fully compensated by mismatched decoding," *IEEE Trans. Inf. Theory*, vol. 69, no. 4, pp. 2152–2164, Apr. 2023

[7] I. Csiszár and J. Körner, *Information Theory: Coding Theorems for Discrete Memoryless Systems*. Cambridge University Press, 2011

[8] R. A. Amjad, "Information rates and error exponents for probabilistic amplitude shaping," in *Proc. IEEE Inf. Theory Workshop (ITW)*, Guangzhou, China, Nov. 2018

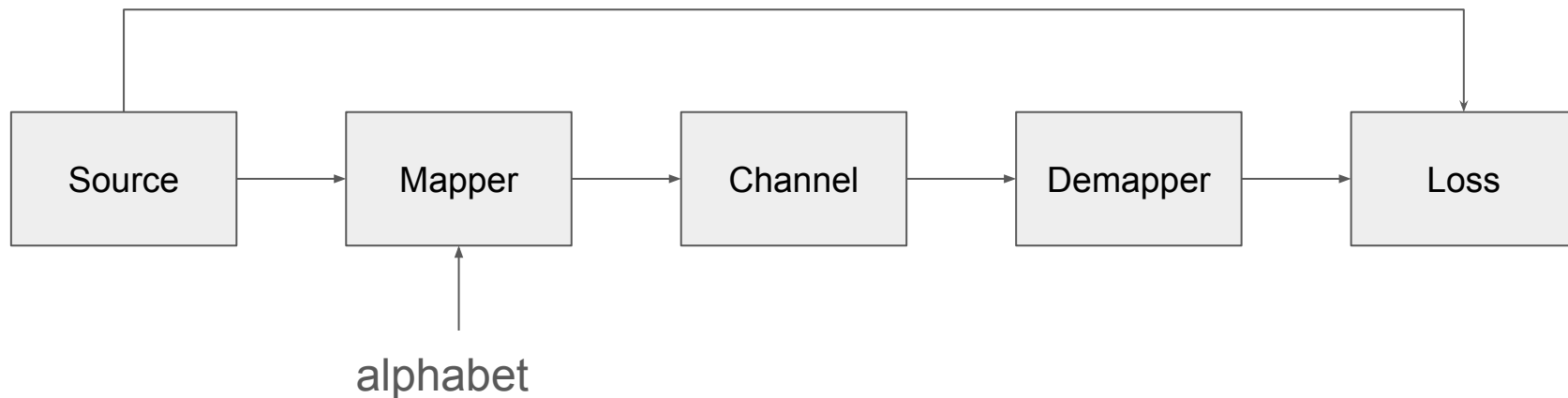


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## Data Driven (“Machine Learning”) Geometric Shaping

- ▶ Geometric shaping is the vanilla case of **data driven** transmitter optimization.
- ▶ Loss function: crossequivocation (“crossentropy”)  $\mathbb{E}[-\log_2 Q(X|Y)] \geq \mathbb{H}(X|Y)$ .
- ▶ Differential channel simulator.
- ▶ Backpropagate loss gradient to alphabet.



## Data Driven Probabilistic Shaping

- ▶ Loss function: crossequivocation is minimized by Dirac delta  $P_X$   
⇒ regularize, e.g., by  $\mathbb{H}(X)$ .
- ▶ How to define a differentiable estimator of  $\mathbb{H}(X)$ ?
- ▶ How to define a differentiable source?

**Problem 3:** Data driven optimization of input distribution  $P_X$

## Remarks on Problem 3

- ▶ For differential entropy, a differentiable estimator exists [9].  
⇒ estimate density in  $x_i$  by distance to closest neighbor.
  - ▶ I use it in [10, Chapter 10] for stochastic channel modelling.
- ▶ There may be no solution, but why?

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[9] J. Jiao, W. Gao, and Y. Han, "The nearest neighbor information estimator is adaptively near minimax rate-optimal," in *Proceedings of the 32nd International Conference on Neural Information Processing Systems*, 2018, pp. 3160–3171

[10] G. Böcherer, *Lecture notes on machine learning for communications*, [Online]. Available: <http://georg-boecherer.de/mlcomm>

# Summary

Practical PAS is well understood, theoretical understanding is still incomplete.

## We lack

- Algorithm for optimizing input distribution **under constrained FEC overhead**.
- Channel coding theorem for **minimum cost DM** (optimal DM?)
- Better understanding difficulty of data-driven input distribution optimization.

## References I

- [1] G. Böcherer, F. Steiner, and P. Schulte, “Bandwidth efficient and rate-matched low-density parity-check coded modulation,” *IEEE Trans. Commun.*, vol. 63, no. 12, pp. 4651–4665, Dec. 2015.
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- [6] J. G. Smith, “The information capacity of amplitude-and variance-constrained scalar Gaussian channels,” *Inf. control*, vol. 18, no. 3, pp. 203–219, 1971.
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## References III

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