



# Grassmannian Product Codebooks for Limited Feedback FD-MIMO

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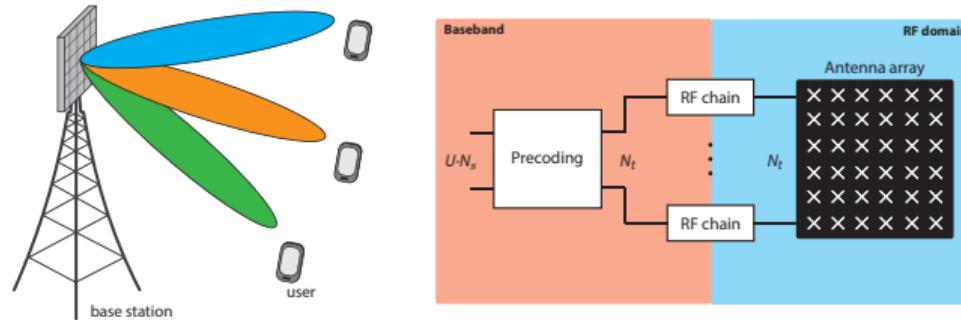
March 12, 2019

### Full-Dimension MIMO Systems

Outer-Tier CSI Feedback

Conclusion

# Full-Dimension MIMO Systems

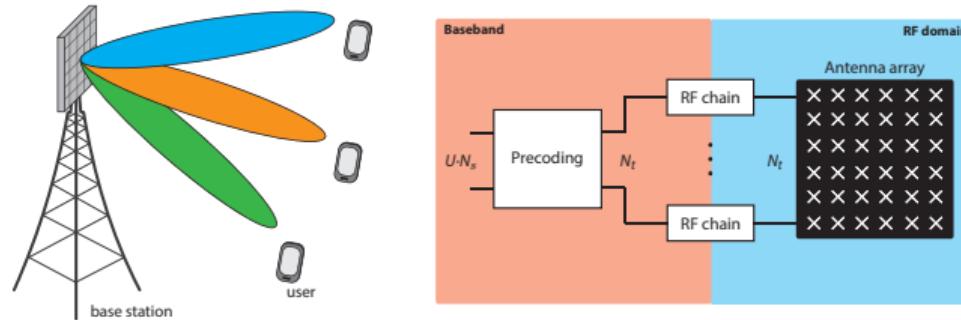


- Utilize two-dimensional active antenna arrays to enable 2D-beamforming (azimuth, elevation)
- Allows to improve spectral efficiency, energy efficiency and network capacity
- We consider downlink linear multi-user MIMO transmission

$$\mathbf{y}_u = \mathbf{H}_u^H \mathbf{F}_u \mathbf{x}_u + \mathbf{H}_u^H \sum_{j=1, j \neq u}^U \mathbf{F}_j \mathbf{x}_j + \mathbf{z}_u$$

- $\mathbf{H}_u \in \mathbb{C}^{N_t \times N_r}$ ,  $\mathbf{F}_u \in \mathbb{C}^{N_t \times N_s}$ :  $N_t$ ,  $N_r$  transmit and receive antennas,  $N_s$  streams per user

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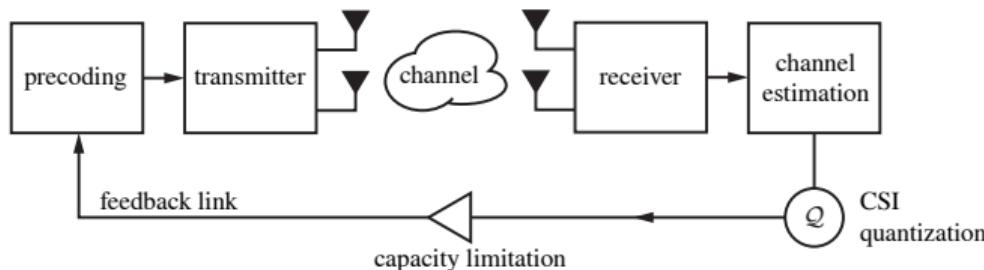


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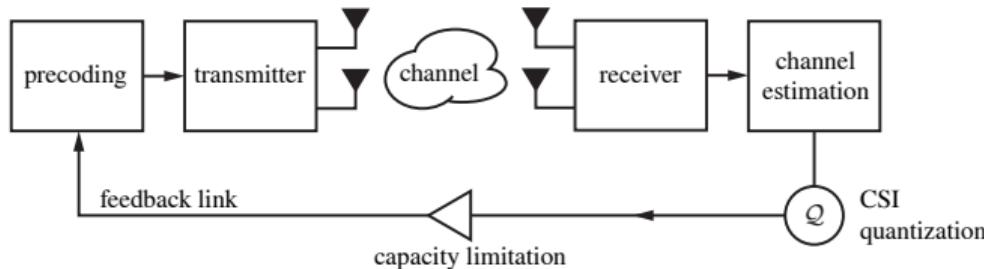
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# Limited Feedback Operation in FDD Systems



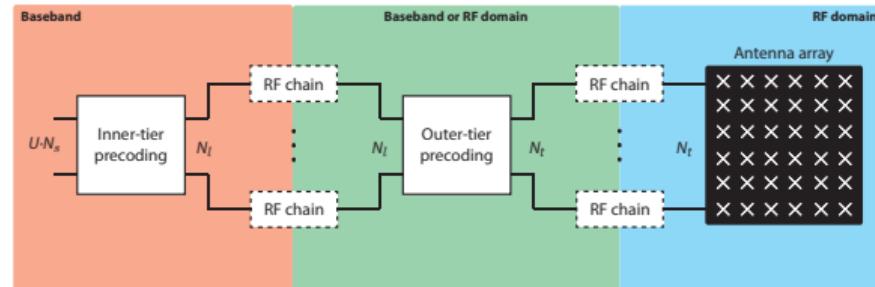
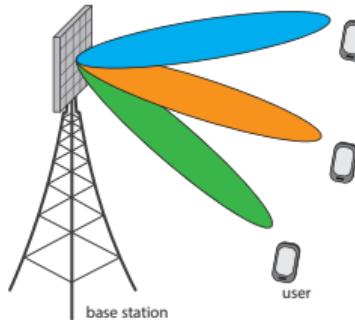
- Linear precoding requires channel state information (CSI) at the transmitter (CSIT)
- Time division duplex (TDD): uplink channel estimation can be employed
- Frequency division duplex (FDD): explicit limited feedback of CSI from the users
  - ⇒ Downlink channel estimation and CSI quantization/feedback required
  - ⇒ Downlink pilot overhead proportional to  $N_t$
  - ⇒ Uplink CSI overhead proportional to  $N_t \cdot N_r$
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## Two-Tier Precoding



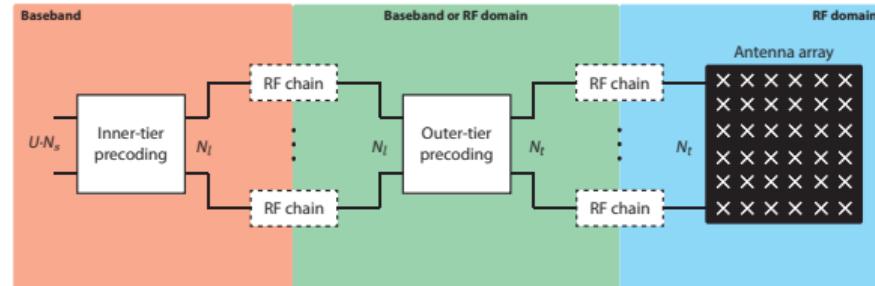
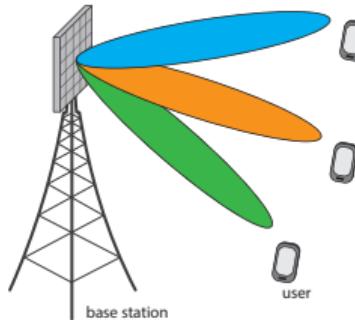
- Split-up precoding into two-tiers:

$$\mathbf{F}_u = \mathbf{F}_o \mathbf{F}_{i,u}$$

1. User-group-specific outer precoding  $\mathbf{F}_o \in \mathbb{C}^{N_t \times N_\ell}$
2. User-specific inner precoding  $\mathbf{F}_{i,u} \in \mathbb{C}^{N_\ell \times N_s}$

- $\mathbf{F}_o$  often considered as wideband precoder (adapted based on channel statistics)
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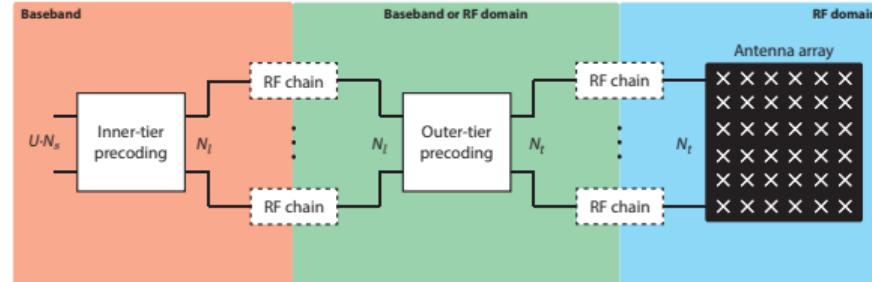
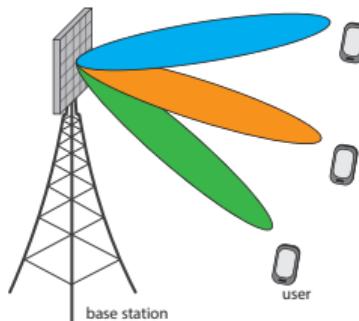
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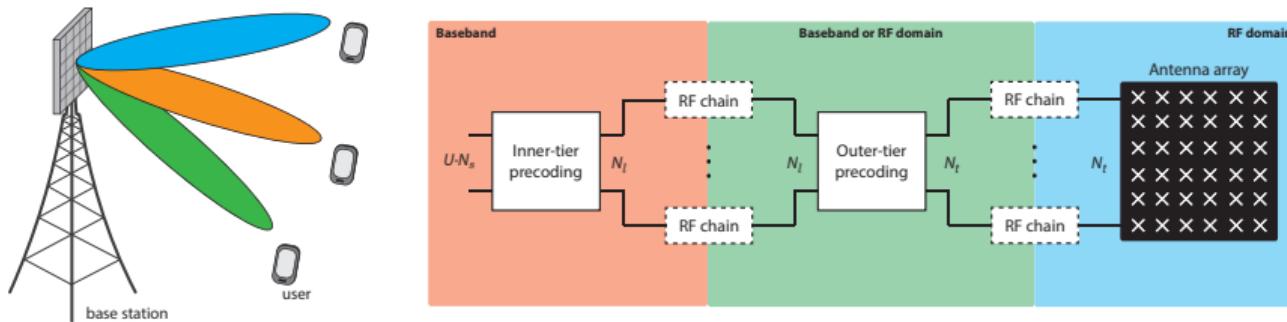
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## Two-Tier Precoding – Advantages/Disadvantages



- Reduced downlink pilot overhead and CSI feedback overhead
  - Infrequent estimation and feedback of the full channel matrix  $\mathbf{H}_u$
  - Per TTI estimation and feedback of  $\mathbf{H}_{o,u} = \mathbf{F}_o^H \mathbf{H}_u \in \mathbb{C}^{N_\ell \times N_r}$ ,  $N_\ell \ll N_t$
- Reduced number of RF chains if  $\mathbf{F}_o$  is implemented in the RF domain (hybrid precoding)
  - Spatial multiplexing capabilities restricted to  $N_\ell$
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## Considered Two-Tier Precoding Strategy

- Outer-tier maximum eigenmode transmission (MET)
  - Maximum eigenmodes of the channel:

$$\mathbf{H}_u \mathbf{H}_u^H = \mathbf{U}_u \boldsymbol{\Sigma}_u \mathbf{U}_u^H,$$
$$\mathbf{U}_u^H \mathbf{U}_u = \mathbf{I}_{N_r}, \quad \boldsymbol{\Sigma}_u = \text{Diag} \left( \sigma_u^{(1)}, \dots, \sigma_u^{(N_r)} \right)$$

- Multi-user MET precoder:

$$\mathbf{F}_o = \left[ \mathbf{U}_1^{(N_s)}, \mathbf{U}_2^{(N_s)}, \dots, \mathbf{U}_U^{(N_s)} \right],$$
$$\mathbf{U}_j^{(N_s)} = \left[ \mathbf{u}_j^{(1)}, \dots, \mathbf{u}_j^{(N_s)} \right]$$

- Inner-tier block-diagonalization (BD) precoding

$$\left( \mathbf{U}_{o,j}^{(N_s)} \right)^H \mathbf{F}_{i,u} \stackrel{!}{=} \mathbf{0}, \quad \forall j \neq u,$$
$$\text{rank} \left( \left( \mathbf{U}_{o,j}^{(N_s)} \right)^H \mathbf{F}_{i,u} \right) = N_s$$

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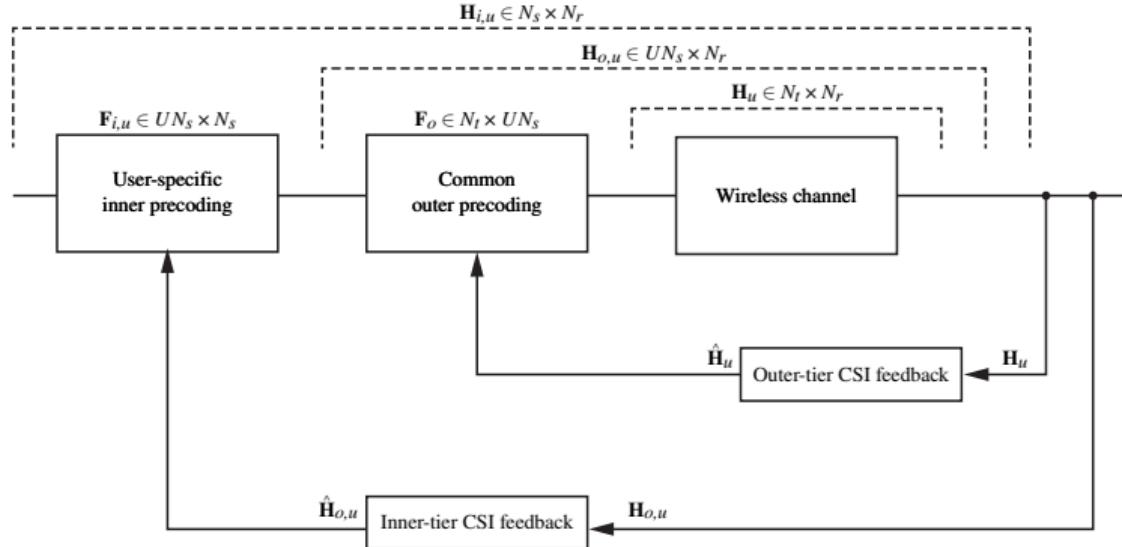
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Full-Dimension MIMO Systems

**Outer-Tier CSI Feedback**

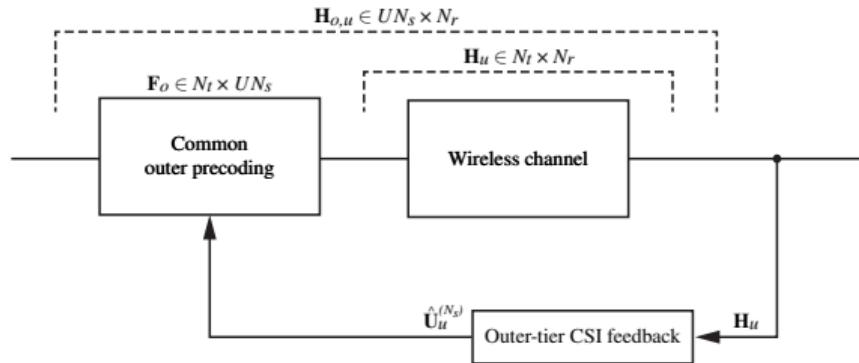
Conclusion

## Two-Tier CSI Feedback



- Many efficient CSI feedback methods are available for small-scale MIMO systems
- These can be employed for the quantization and feedback of  $\mathbf{H}_{o,u}$   
⇒ We focus on computationally and rate-distortion efficient outer-tier CSI feedback of  $\mathbf{H}_u$

## Outer-Tier CSI Feedback



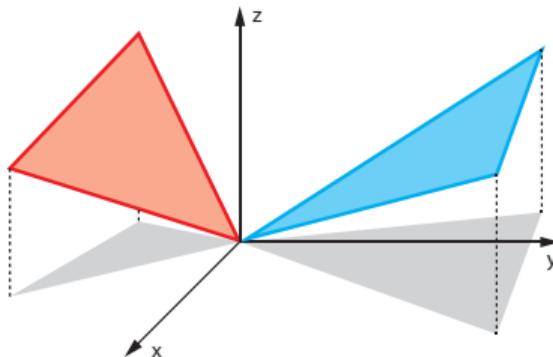
- What information does the transmitter need about  $\mathbf{H}_u$  to enable MET precoding?

$$\mathbf{F}_o = \left[ \mathbf{U}_1^{(N_s)}, \dots, \mathbf{U}_U^{(N_s)} \right] \equiv \left[ \mathbf{U}_1^{(N_s)} \mathbf{Q}_1, \dots, \mathbf{U}_U^{(N_s)} \mathbf{Q}_U \right],$$
$$\mathbf{Q}_i^H \mathbf{Q}_i = \mathbf{Q}_i \mathbf{Q}_i^H = \mathbf{I}_{N_s}$$

⇒ Subspace information is sufficient, since we do not consider power-loading over the eigenmodes

⇒ We apply Grassmannian quantization to span  $(\mathbf{U}_1^{(N_s)})$  on the Grassmann manifold  $\mathcal{G}_{N_s}^{(N_t)}$

## Grassmann Manifold Basics

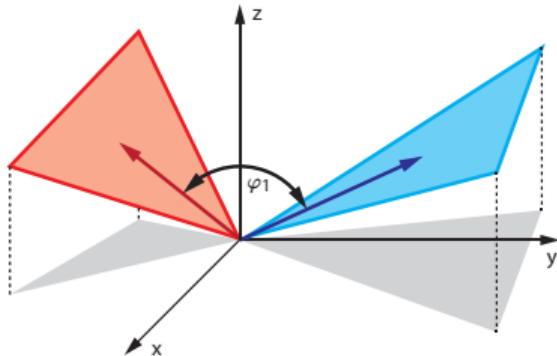


- $\mathcal{G}_{N_s}^{(N_t)}$  manifold of  $N_s$ -dimensional subspaces of the  $N_t$ -dimensional real/complex Euclidean space
- Chordal distance between two subspaces defined by their orthogonal bases  $\mathbf{U}_1, \mathbf{U}_2 \in \mathbb{C}^{N_t \times N_s}$

$$d_{\mathcal{G}}^2(\mathbf{U}_1, \mathbf{U}_2) = N_s - \text{tr} \left( \mathbf{U}_1^H \mathbf{U}_2 \mathbf{U}_2^H \mathbf{U}_1 \right) = \sum_{i=1}^{N_s} \sin \varphi_i^2$$

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## Single-Stage Outer-Tier CSI Feedback

- The SNR-loss of MET-BD precoding with imperfect CSIT w.r.t. perfect CSIT is determined by the chordal distance CSI error  $\Rightarrow$  minimum chordal distance quantization:

$$\hat{\mathbf{U}}_{j,\text{single}}^{(N_s)} = \arg \min_{\mathbf{W}_\ell \in \mathcal{Q}_{N_s}^{(N_t)}} d_{\mathcal{G}}^2 \left( \mathbf{U}_j^{(N_s)}, \mathbf{W}_\ell \right)$$

- Quantization codebook  $\mathcal{Q}_{N_s}^{(N_t)} = \{ \mathbf{W}_\ell \in \mathbb{C}^{N_t \times N_s} | \mathbf{W}_\ell^H \mathbf{W}_\ell = \mathbf{I}_{N_s}, \forall \ell \}$  of size  $2^b$
- Rayleigh fading channels: random subspace quantization is asymptotically efficient

$$\bar{d}_{c,\text{single}}^2 = \mathbb{E} \left( d_{\mathcal{G}}^2 \left( \mathbf{U}_j^{(N_s)}, \hat{\mathbf{U}}_{j,\text{single}}^{(N_s)} \right) \right) \propto 2^{-b/(N_s(N_t - N_s))}$$

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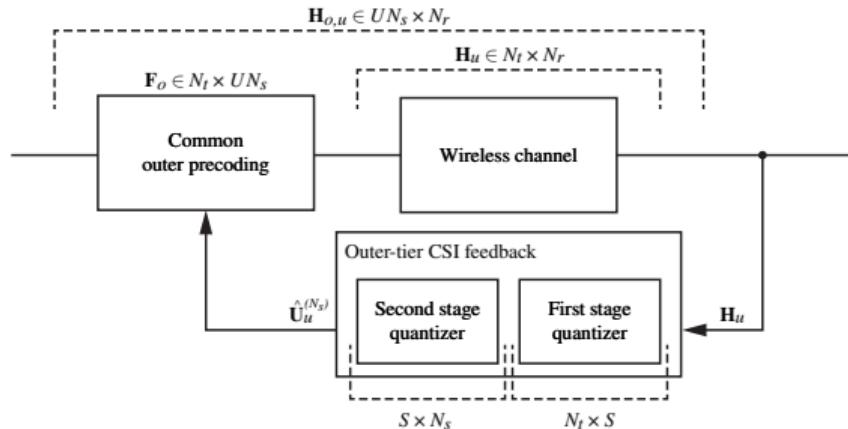
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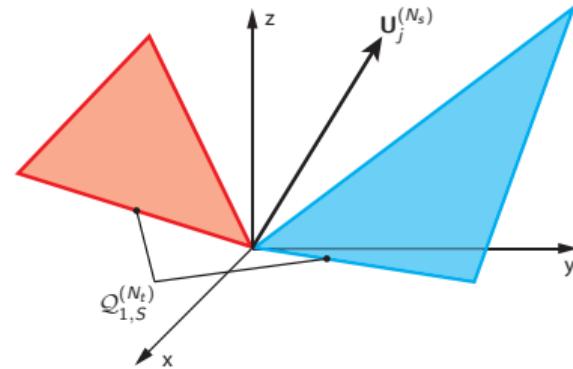
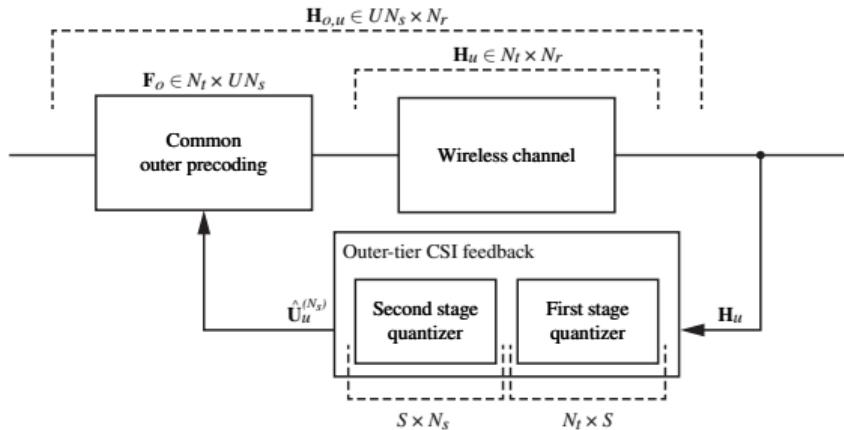


- Employ a product-codebook construction to reduce the quantization complexity

$$\hat{\mathbf{U}}_1 \in \mathcal{Q}_{1,S}^{(N_t)} \subset \mathbb{C}^{N_t \times S}, \quad \hat{\mathbf{U}}_2 \in \mathcal{Q}_{2,N_s}^{(S)} \subset \mathbb{C}^{S \times N_s}$$

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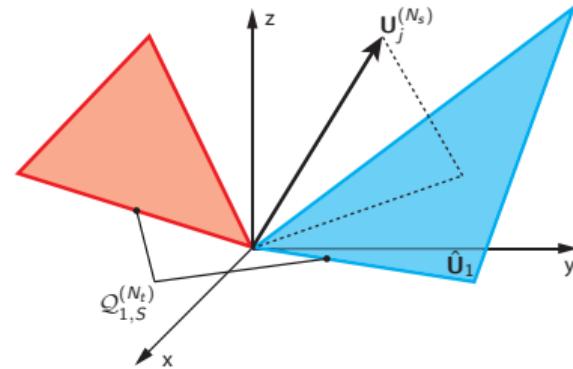
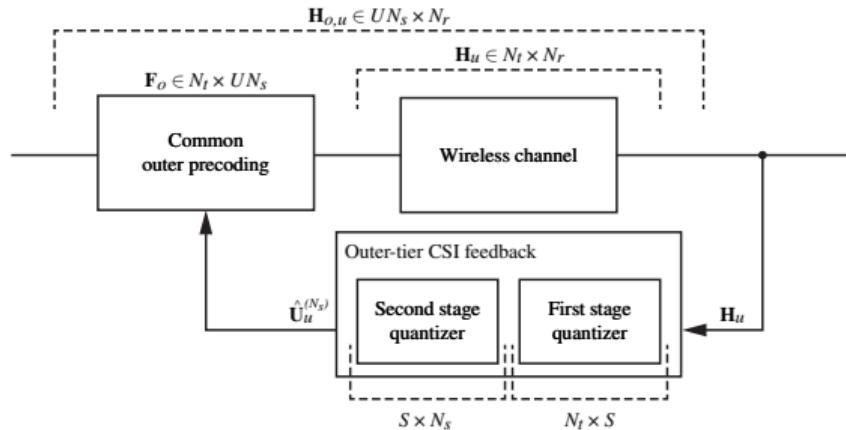


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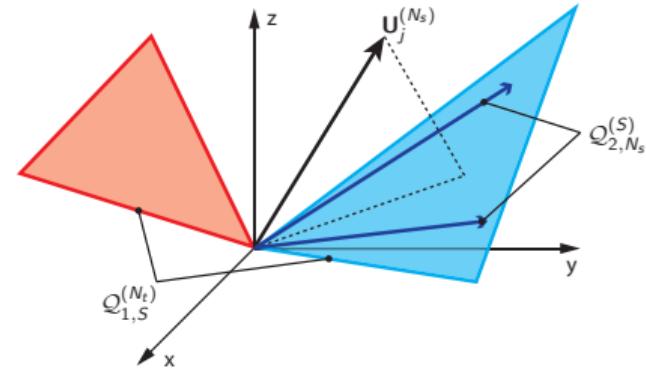
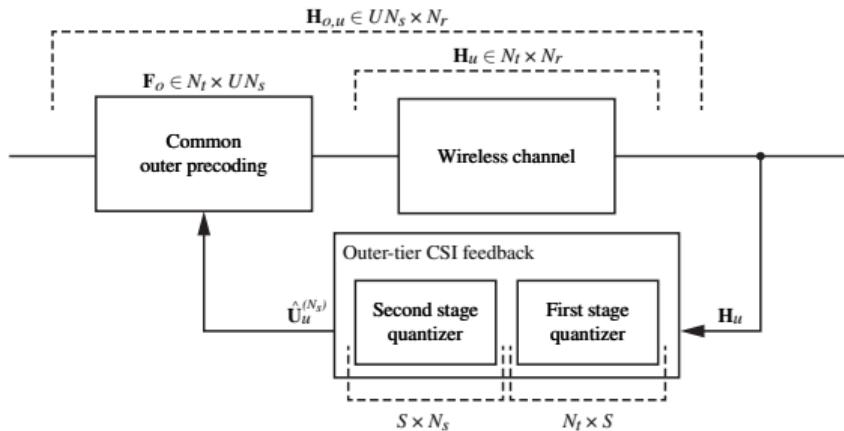


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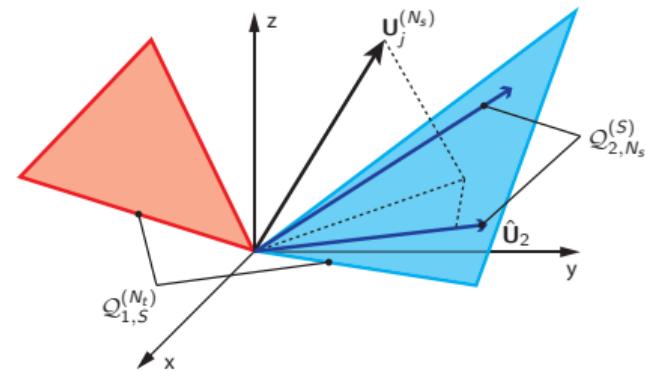
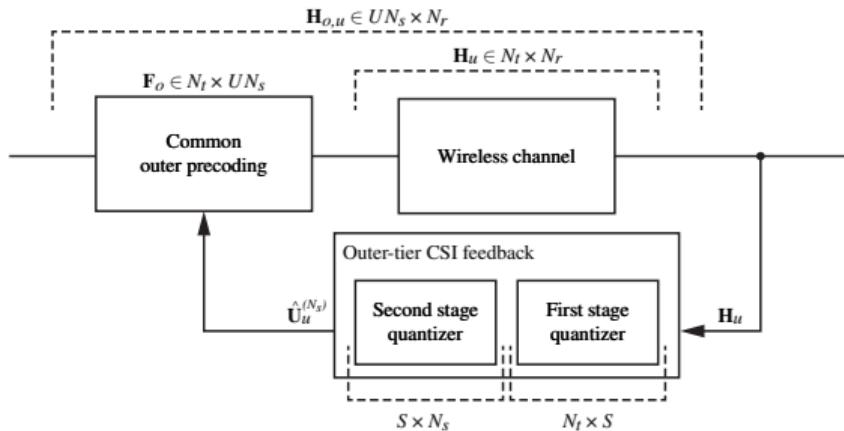


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## Multi-Stage Outer-Tier CSI Feedback (II)

- First-stage quantizer:

$$\hat{\mathbf{U}}_1 = \underset{\mathbf{W}_\ell \in \mathcal{Q}_{1,S}^{(N_t)}}{\arg \min} d_G^2 \left( \mathbf{U}_j^{(N_s)}, \mathbf{W}_\ell \right) \in \mathbb{C}^{N_t \times S}$$

- Subspace-quantization based combining

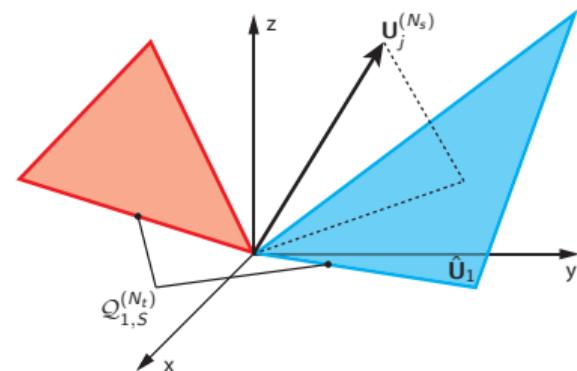
$$\mathbf{B} = \hat{\mathbf{U}}_1^H \mathbf{U}_j^{(N_s)} \left( \left( \mathbf{U}_j^{(N_s)} \right)^H \hat{\mathbf{U}}_1 \hat{\mathbf{U}}_1^H \mathbf{U}_j^{(N_s)} \right)^{-\frac{1}{2}} \in \mathbb{C}^{S \times N_s},$$

$$\mathbf{B}^H \mathbf{B} = \mathbf{I}_{N_s}, \quad d_G^2 \left( \hat{\mathbf{U}}_1, \mathbf{U}_j^{(N_s)} \right) = d_G^2 \left( \hat{\mathbf{U}}_1 \mathbf{B}, \mathbf{U}_j^{(N_s)} \right)$$

- Second-stage quantizer:

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- Total quantized CSI:  $\hat{\mathbf{U}}_{j,\text{dual}}^{(N_s)} = \hat{\mathbf{U}}_1 \hat{\mathbf{U}}_2 \in \mathbb{C}^{N_t \times N_s}$



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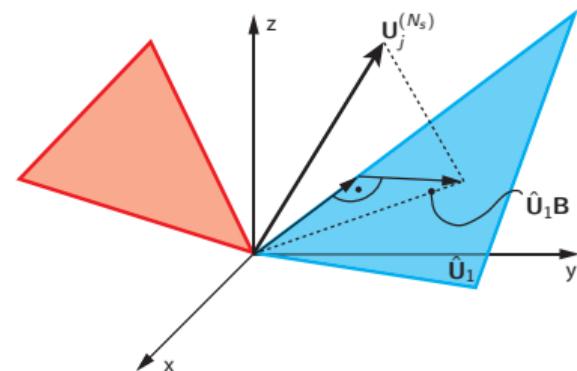
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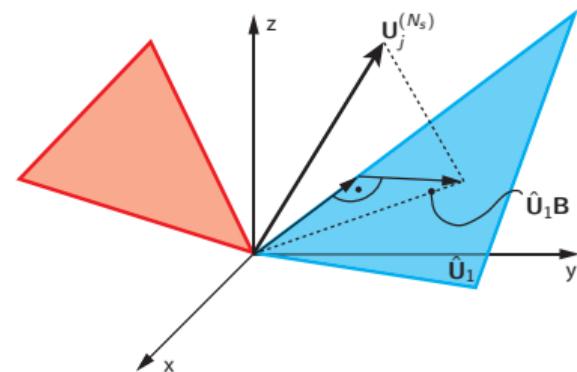
$$\mathbf{B} = \hat{\mathbf{U}}_1^H \mathbf{U}_j^{(N_s)} \left( \left( \mathbf{U}_j^{(N_s)} \right)^H \hat{\mathbf{U}}_1 \hat{\mathbf{U}}_1^H \mathbf{U}_j^{(N_s)} \right)^{-\frac{1}{2}} \in \mathbb{C}^{S \times N_s},$$

$$\mathbf{B}^H \mathbf{B} = \mathbf{I}_{N_s}, \quad d_G^2 \left( \hat{\mathbf{U}}_1, \mathbf{U}_j^{(N_s)} \right) = d_G^2 \left( \hat{\mathbf{U}}_1 \mathbf{B}, \mathbf{U}_j^{(N_s)} \right)$$

- Second-stage quantizer:

$$\hat{\mathbf{U}}_2 = \arg \min_{\mathbf{W}_\ell \in \mathcal{Q}_{2,N_s}^{(S)}} d_G^2 \left( \mathbf{B}, \mathbf{W}_\ell \right)$$

- Total quantized CSI:  $\hat{\mathbf{U}}_{j,\text{dual}}^{(N_s)} = \hat{\mathbf{U}}_1 \hat{\mathbf{U}}_2 \in \mathbb{C}^{N_t \times N_s}$



## Multi-Stage Outer-Tier CSI Feedback (II)

- First-stage quantizer:

$$\hat{\mathbf{U}}_1 = \arg \min_{\mathbf{W}_\ell \in \mathcal{Q}_{1,S}^{(N_t)}} d_G^2 \left( \mathbf{U}_j^{(N_s)}, \mathbf{W}_\ell \right) \in \mathbb{C}^{N_t \times S}$$

- Subspace-quantization based combining

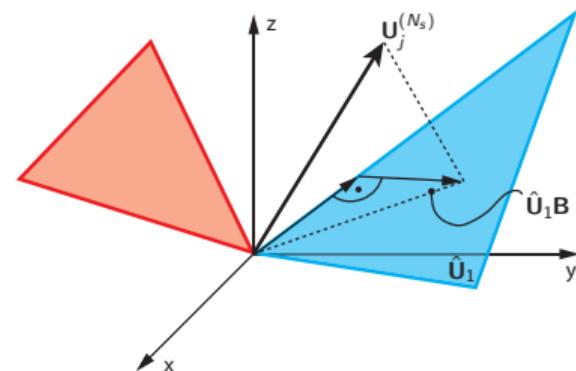
$$\mathbf{B} = \hat{\mathbf{U}}_1^H \mathbf{U}_j^{(N_s)} \left( \left( \mathbf{U}_j^{(N_s)} \right)^H \hat{\mathbf{U}}_1 \hat{\mathbf{U}}_1^H \mathbf{U}_j^{(N_s)} \right)^{-\frac{1}{2}} \in \mathbb{C}^{S \times N_s},$$

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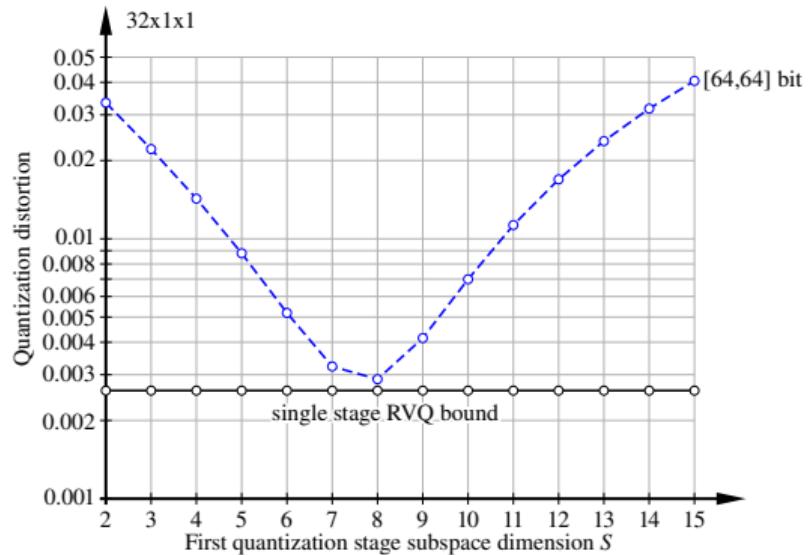
- Second-stage quantizer:

$$\hat{\mathbf{U}}_2 = \arg \min_{\mathbf{W}_\ell \in \mathcal{Q}_{2,N_s}^{(S)}} d_G^2 \left( \mathbf{B}, \mathbf{W}_\ell \right)$$

- Total quantized CSI:  $\hat{\mathbf{U}}_{j,\text{dual}}^{(N_s)} = \hat{\mathbf{U}}_1 \hat{\mathbf{U}}_2 \in \mathbb{C}^{N_t \times N_s}$



## Single-Stage vs. Multi-Stage Outer-Tier CSI Feedback – Rayleigh Fading

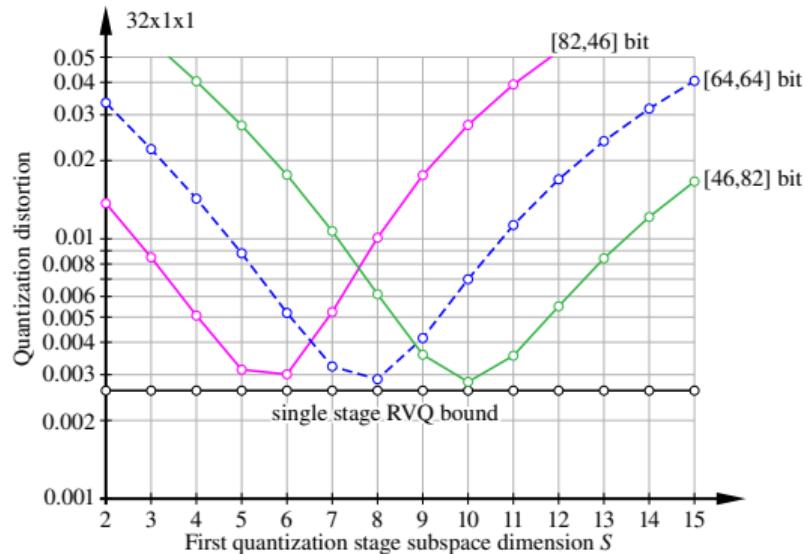


- The product codebook quantization error is determined by:

$$\bar{d}_{c,\text{dual}}^2 = \mathbb{E} \left( d_{\mathcal{G}}^2 \left( \mathbf{U}_j^{(N_s)}, \hat{\mathbf{U}}_{j,\text{dual}}^{(N_s)} \right) \right) = N_s - \frac{1}{N_s} (N_s - \bar{d}_{c,1}^2) (N_s - \bar{d}_{c,2}^2)$$

- Notice  $2^{64} + 2^{64} = 2^{65} \ll 2^{128} \rightarrow$  significant complexity reduction

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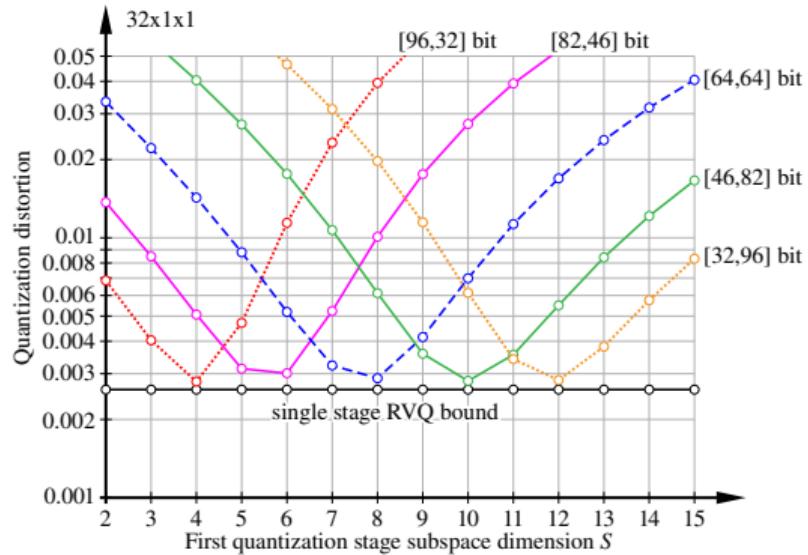


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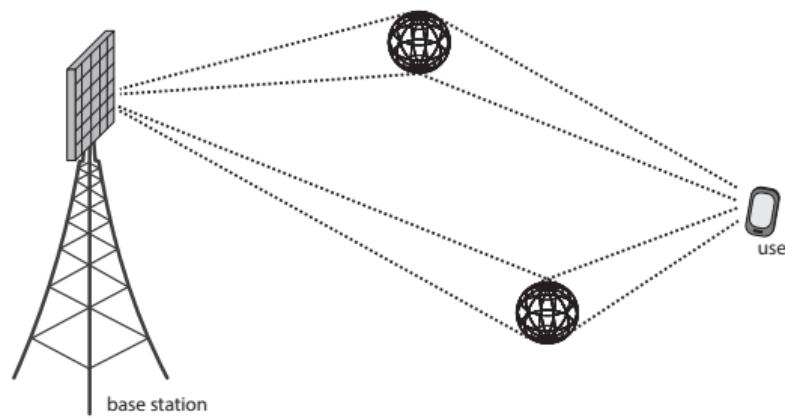


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# Multi-Scattering Directional Channels



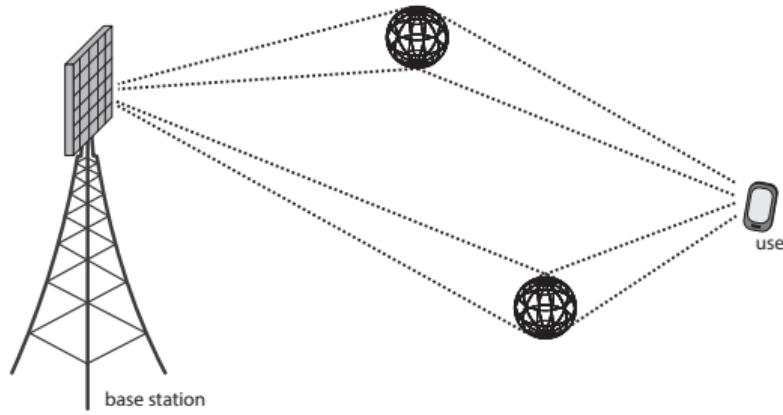
- Multi-scattering directional channel model:

$$\mathbf{H}_u = \sqrt{N_t N_r} \sum_{p=1}^{N_p} \alpha_{p,u} \mathbf{a}_t(\phi_{p,u}) \mathbf{a}_r(\theta_{p,u})^T$$

- Antenna array response vector of a uniform linear array:

$$[\mathbf{a}_t^{\text{ULA}}(\phi)]_k = \frac{g_e(\phi)}{\sqrt{N_t}} \exp\left(j \frac{2\pi d}{\lambda} (k-1) \sin(\phi)\right)$$

# Multi-Scattering Directional Channels



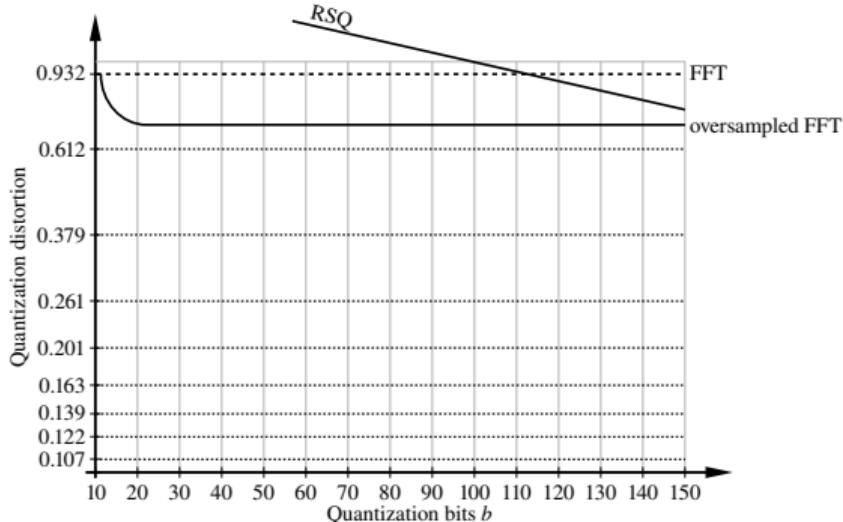
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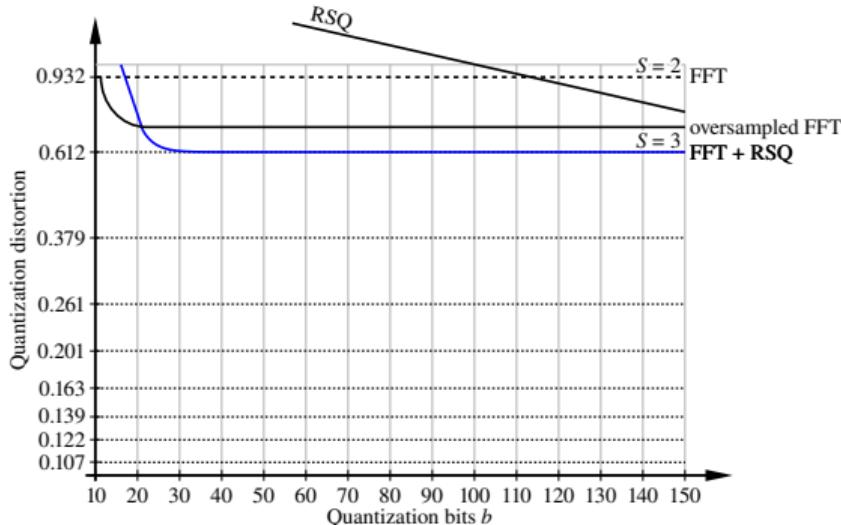


- Multi-scattering directional channels: DFT-based codebooks perform well for small  $b$

$$\mathcal{Q}_{S,\text{DFT}}^{(N_t, N_{\text{DFT}})} = \left\{ \frac{1}{\sqrt{N_t}} [\mathbf{D}_{N_{\text{DFT}}}]_{1:N_t, S} \mid \forall S, |S| = S \right\}, \quad [\mathbf{D}_{N_{\text{DFT}}}]_{\ell, k} = e^{-j \frac{2\pi(\ell-1)(k-1)}{N_{\text{DFT}}}}$$

- Oversampled DFT codebooks exhibit an error saturation for  $N_p \geq 1$  due to unit-modulus constraint
- Example:  $N_t = 64$ ,  $N_r = N_s = 2$ ,  $N_p = 4$

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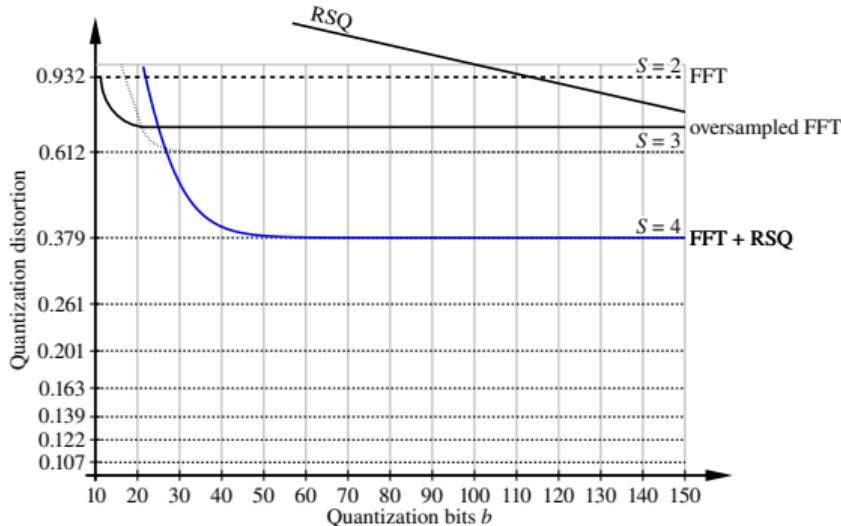


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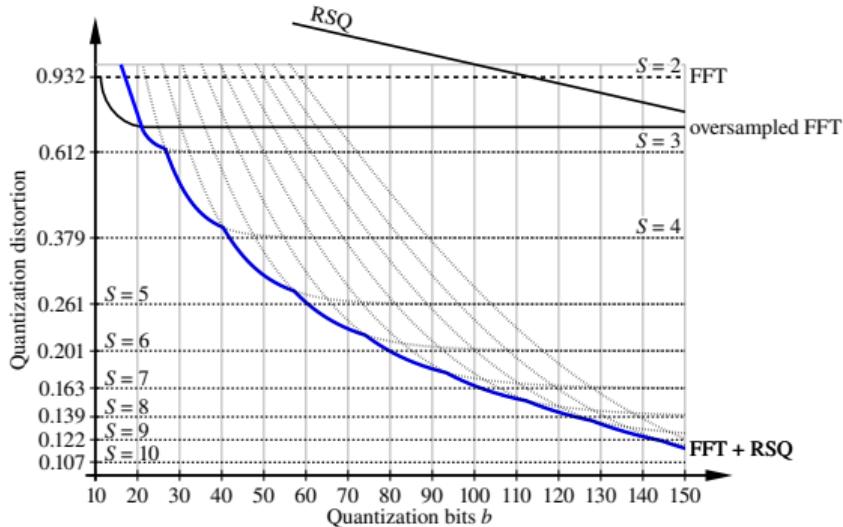


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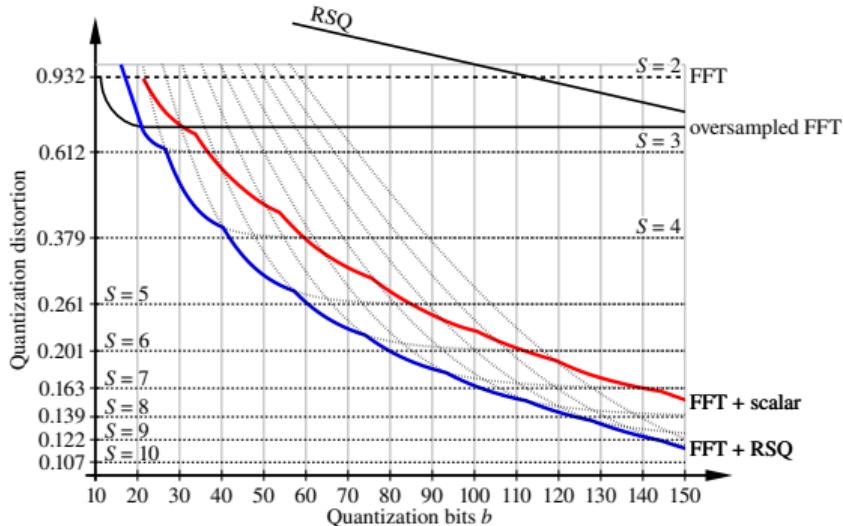


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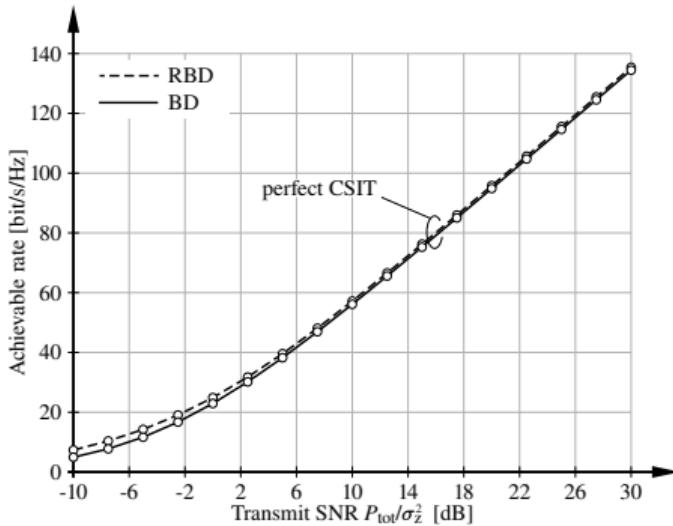
## Throughput Performance



- Measured channel traces of Nokia Bell Labs, Stuttgart
- $N_t = 64$ ,  $N_r = N_s = 2$ ,  $U = 6$  users with velocities between 15 and 25 km/h @2 GHz center frequency

$$\left( \log_2 \left( \binom{N_{\text{DFT}}}{S} \right) + S N_s b_s \right) R_f R_t \approx 5.36 \text{ bit/ms/MHz.}$$

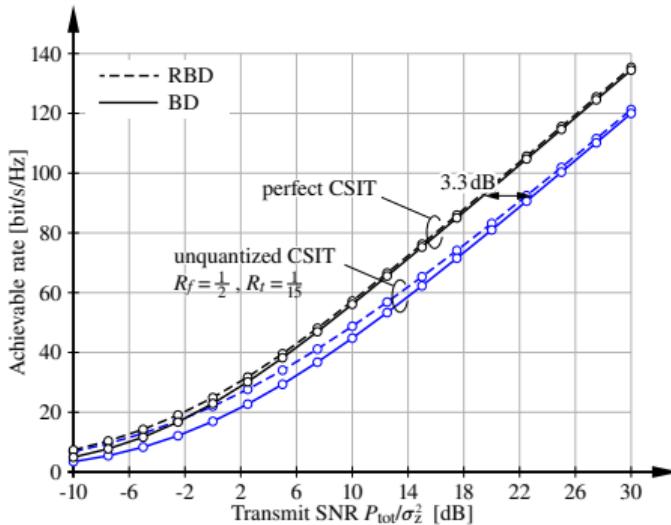
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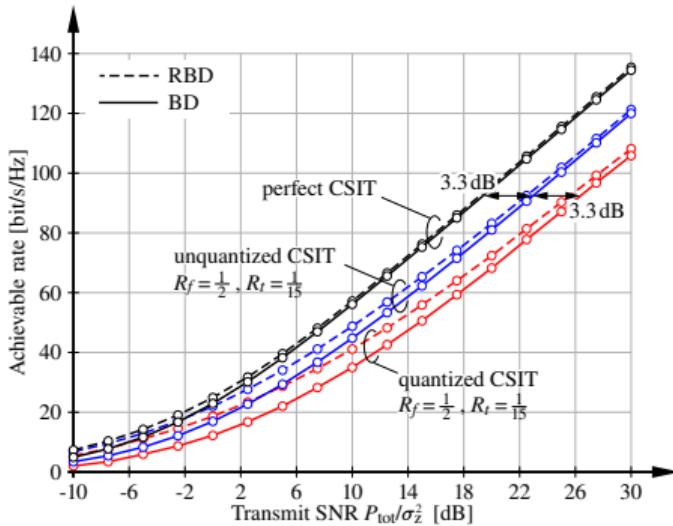
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## Contents

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**Conclusion**

## Summary and Conclusion

- Grassmannian dual-stage product codebooks can perform very close to single-stage quantization when employing RVQ/RSQ
- Significant complexity reduction by utilizing two relatively small codebooks rather than a single large codebook
- In multi-scattering channels, the product codebook can additionally mitigate the error floor of DFT codebooks
- Future work: further extension to multi-stage product codebooks?